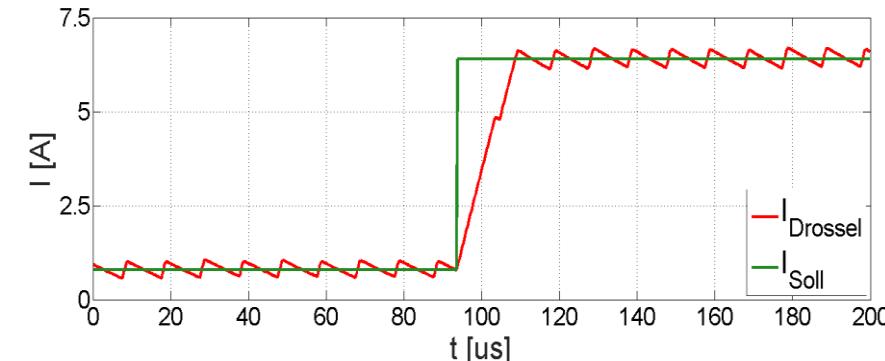


# An Easily Adaptive Current Control Method for DC-DC Converters

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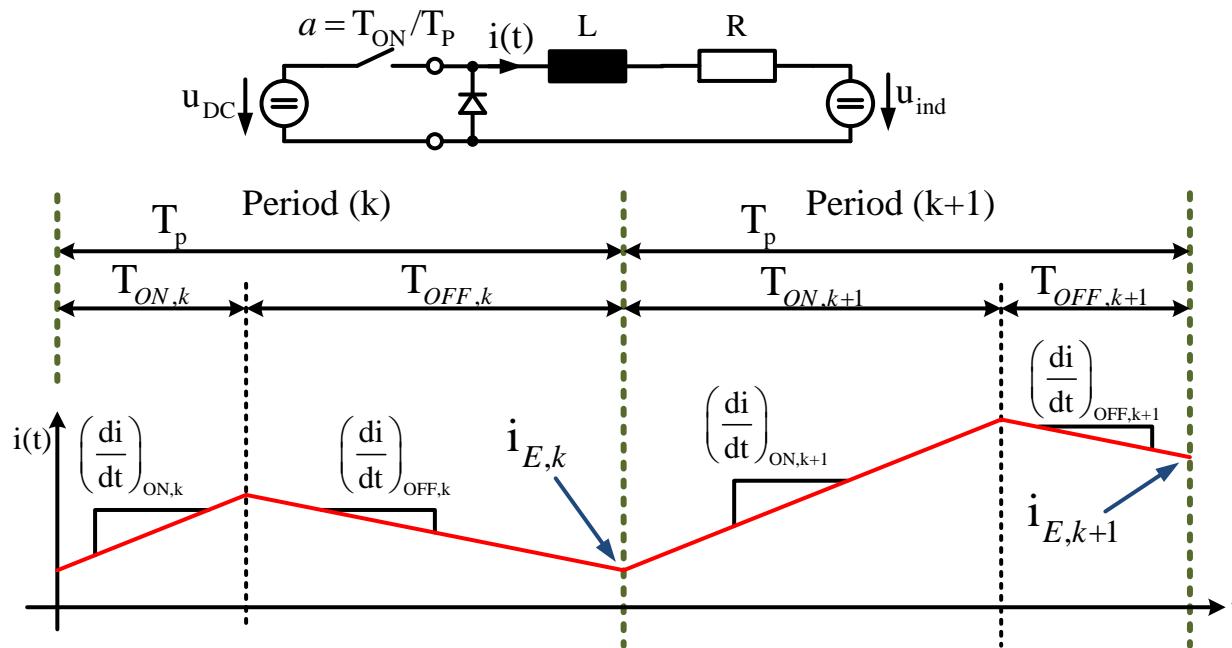
II. Easy Current Slope Detection (ECSD)

III. Implementing the ECSD with the DACC

IV. Conclusion

# I. Direct Adaptive Current Control

## Basic Principle



$$i_{E,k+1} = i_{E,k} + T_{ON,k+1} \left( \frac{di}{dt} \right)_{ON,k} + T_{OFF,k+1} \left( \frac{di}{dt} \right)_{OFF,k}$$

with  $u_{DC}, L, R, u_{ind} \approx \text{const.}$  for  $2 \cdot T_p$

Knowledge of the  
**current slopes**,  
depending on each applied  
**switching state**



**Duty cycle**  
for a given current setpoint  
can be calculated.

No control path  
parameters  
required!

# I. Direct Adaptive Current Control

## Features & Challenges

### ***Features*** of the DACC:

- ✓ No control path parameters required
- ✓ Completely adaptive
- ✓ Excellent control quality
- ✓ Dead-Beat characteristics
- ✓ Inherently stable
- ✓ Suitable for all kinds of DC/DC-converters
- ✓ Adaptable for 3-phase machine control



**current slopes  
of each switching state  
must be known**

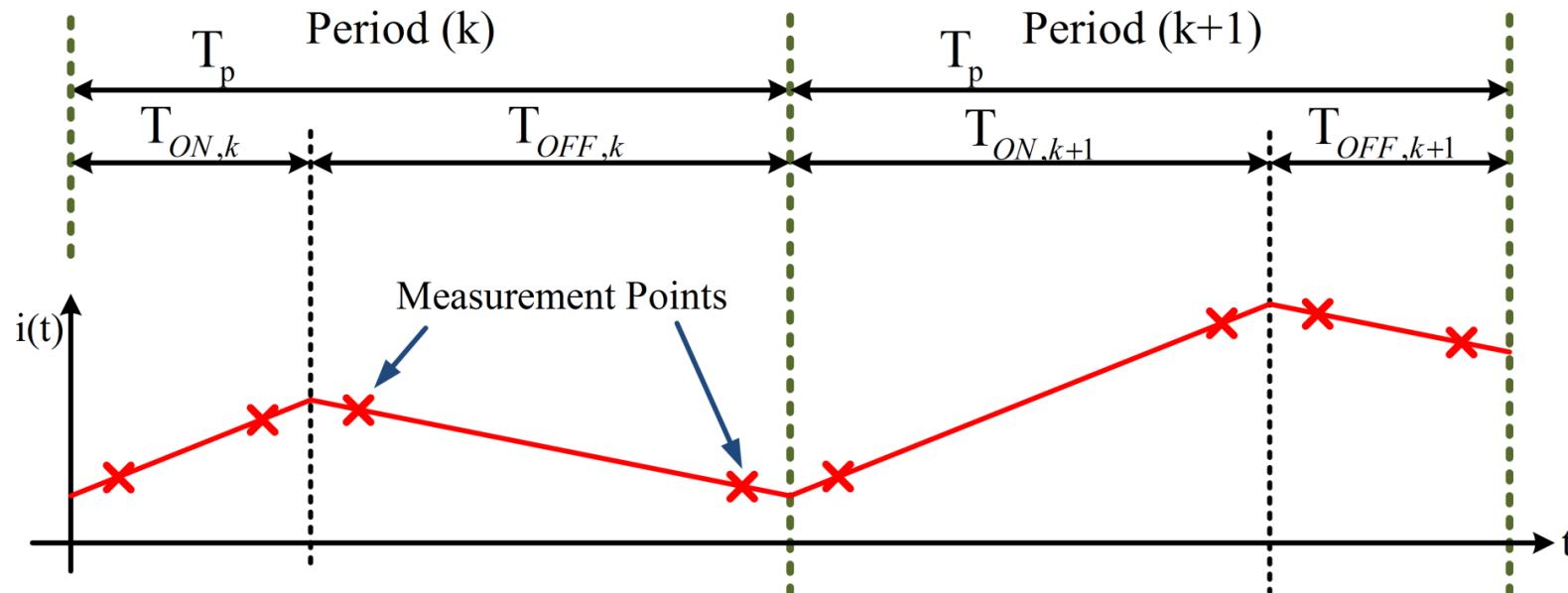
### ***Challenges*** with fast switching DC-DC-Converters:

- short measurement time
- noise
- low-cost demand



## II. Easy Current Slope Detection (ECSD)

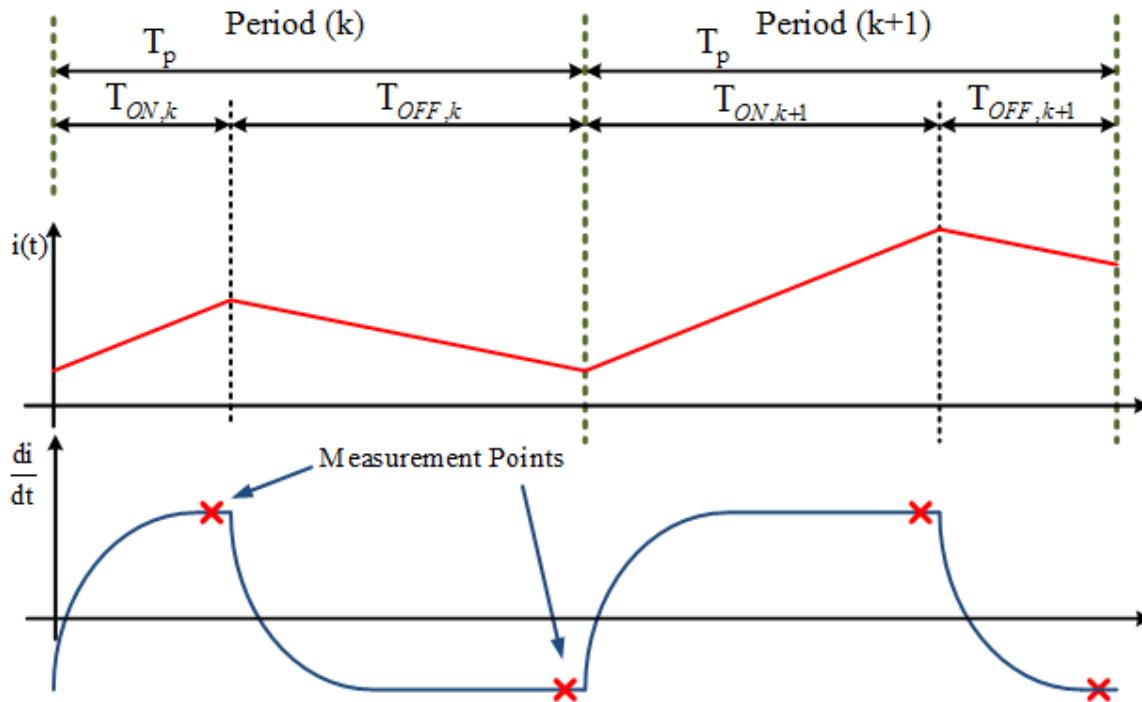
### Measuring the current slopes – „2-point-method“



- ✓ Simple
- ✗ very sensitive to noise
- ✗ difficult with small duty cycles / high switching frequency
- ✗ Special trigger necessary

## II. Easy Current Slope Detection (ECSD)

### Measuring the current slopes – $di/dt$ -sensor

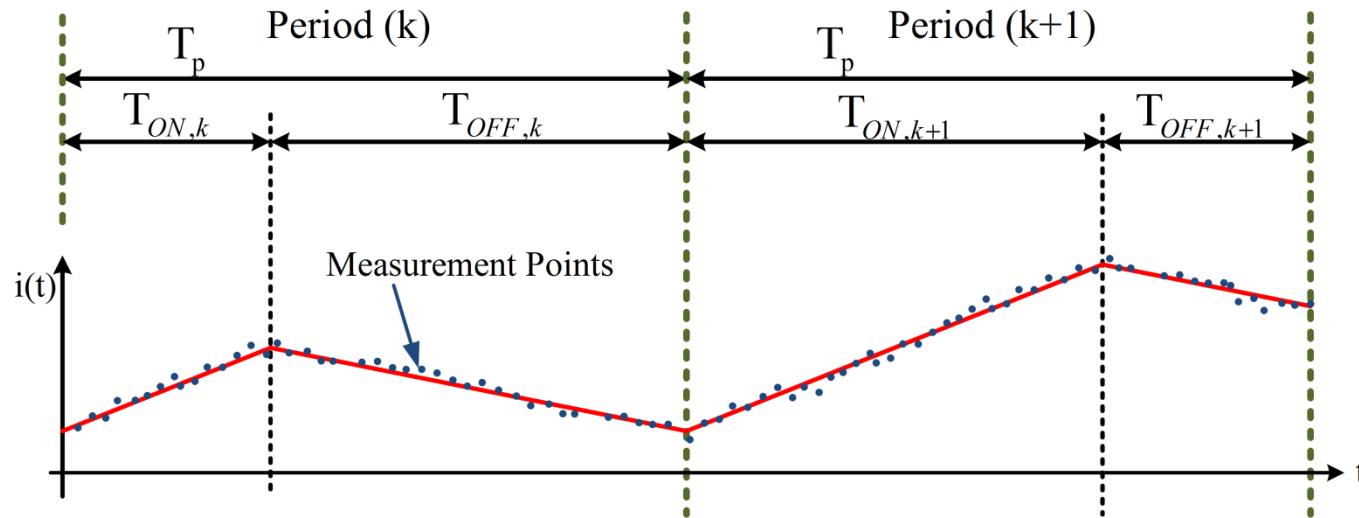


- ✓ direct measurement of the current slopes with a  $di/dt$ -Sensor or analog differentiator circuit
- ✗ special hardware necessary
- ✗ not ideal for high switching frequencies

## II. Easy Current Slope Detection (ECSD)

### Measuring the current slopes – Oversampling

- Measurement with fast oversampling
- Calculation of the current slopes with a Least-Squares-Estimator

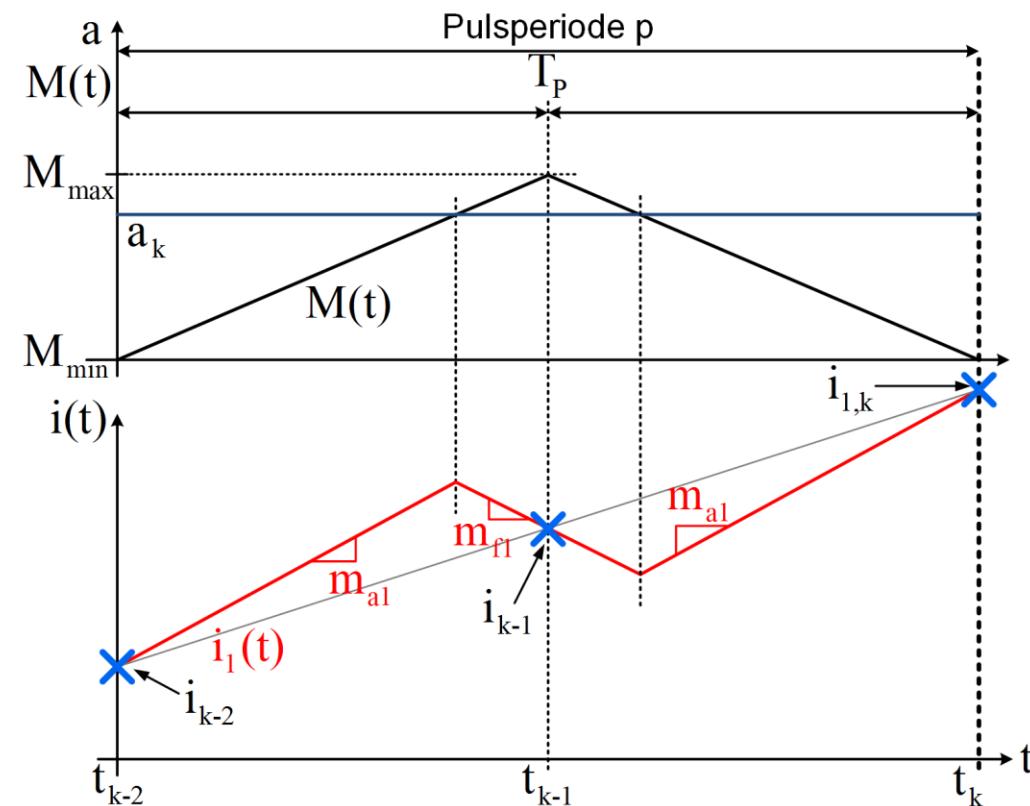


- ✓ very reliable and well-proven
- ✗ Fast AD-Conversion and calculations necessary
- ✗ not suitable for high switching frequencies
- ✗ not low cost

## II. Easy Current Slope Detection (ECSD)

### Standard features in Low-Cost Microcontrollers

- Measuring times:
  - $M(t) = 0$
  - $M(t) = \text{max}$
- Sampling times „farthest away“ from duty cycle transitions



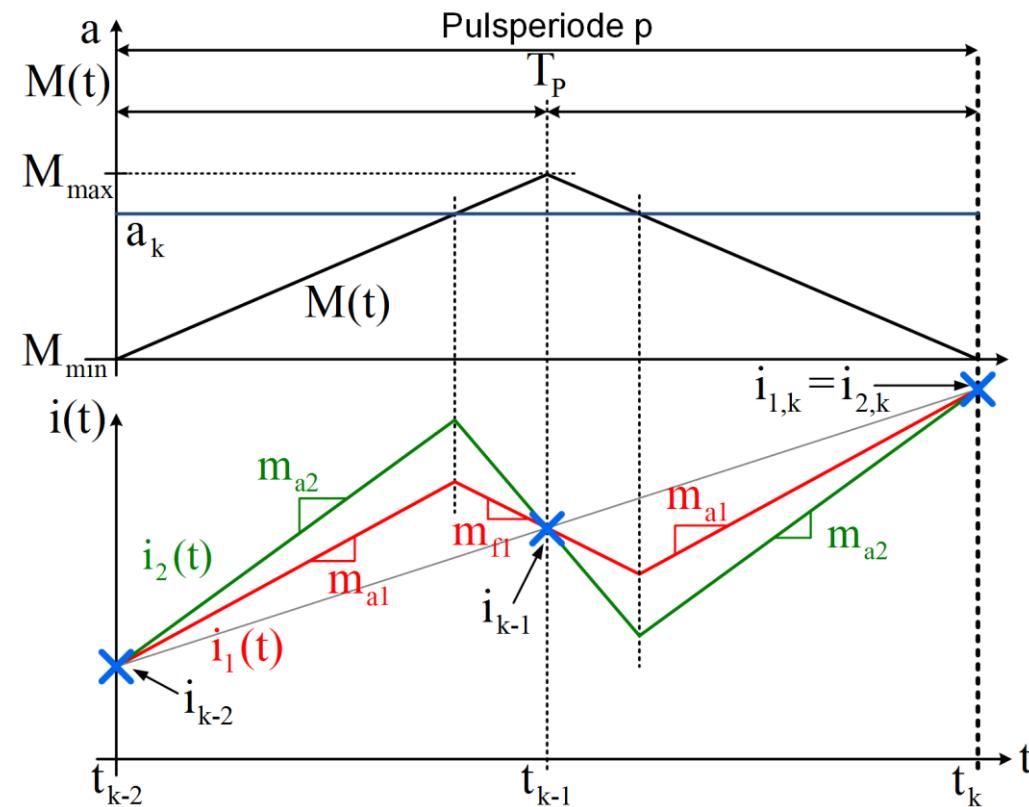
„a“ → active  
„f“ → freewheeling

## II. Easy Current Slope Detection (ECSD)

### Standard features in Low-Cost Microcontrollers

- Measuring times:
  - $M(t) = 0$
  - $M(t) = \text{max}$
- Sampling times „farthest away“ from duty cycle transitions
- Different control paths
  - different current slopes
  - Identical measurement values

→ Current slopes not detectable!



„a“ → active  
„f“ → freewheeling

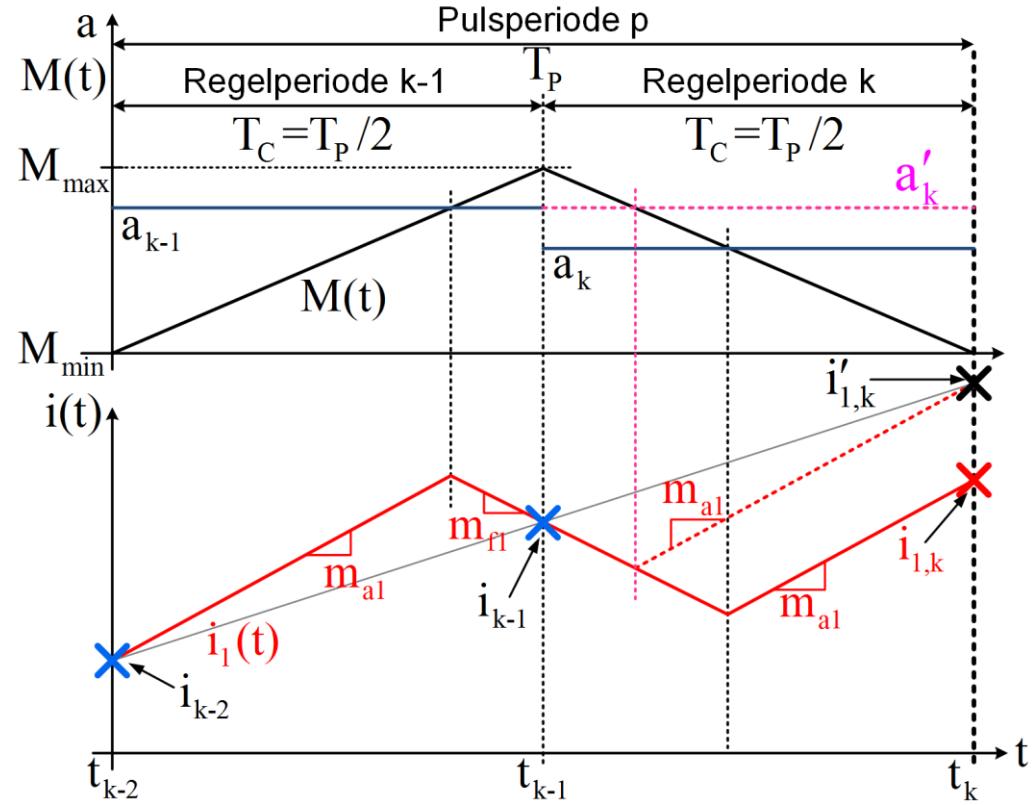
## II. Easy Current Slope Detection (ECSD)

### „The Trick“

*Trick:*

Variation of subsequent duty-cycles (Jitter)

$$a_{k-1} \neq a_k$$



## II. Easy Current Slope Detection (ECSD)

### „The Trick“

*Trick:*

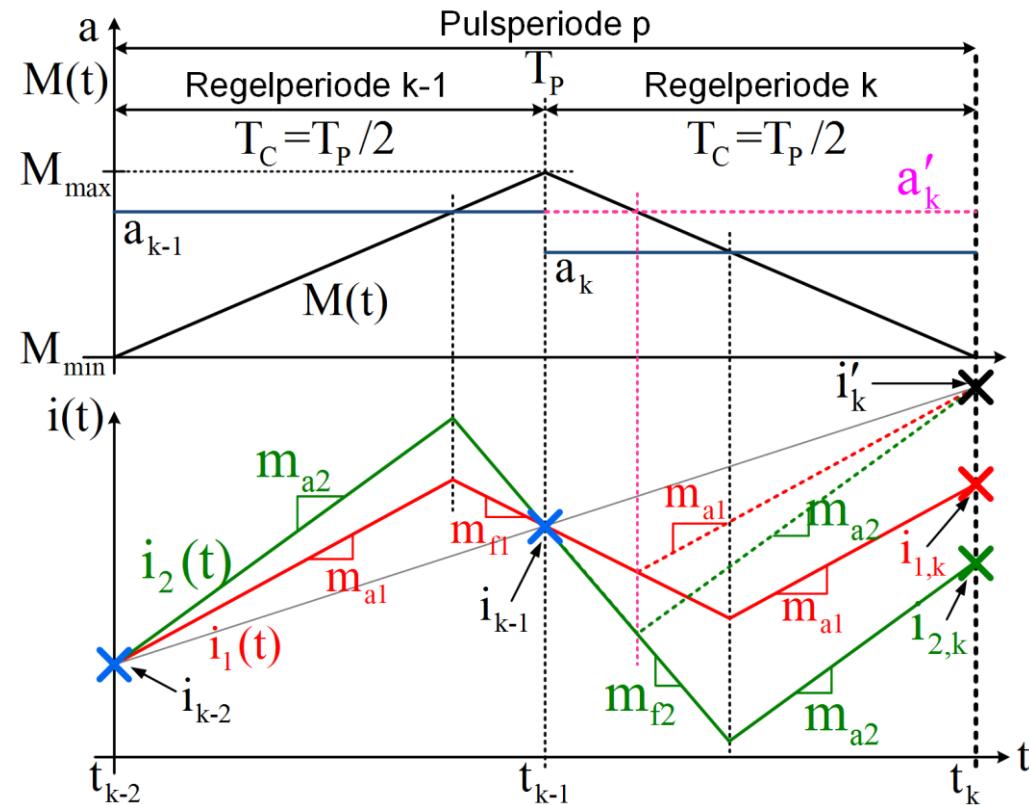
Variation of subsequent duty-cycles (Jitter)

$$a_{k-1} \neq a_k$$

Different control paths

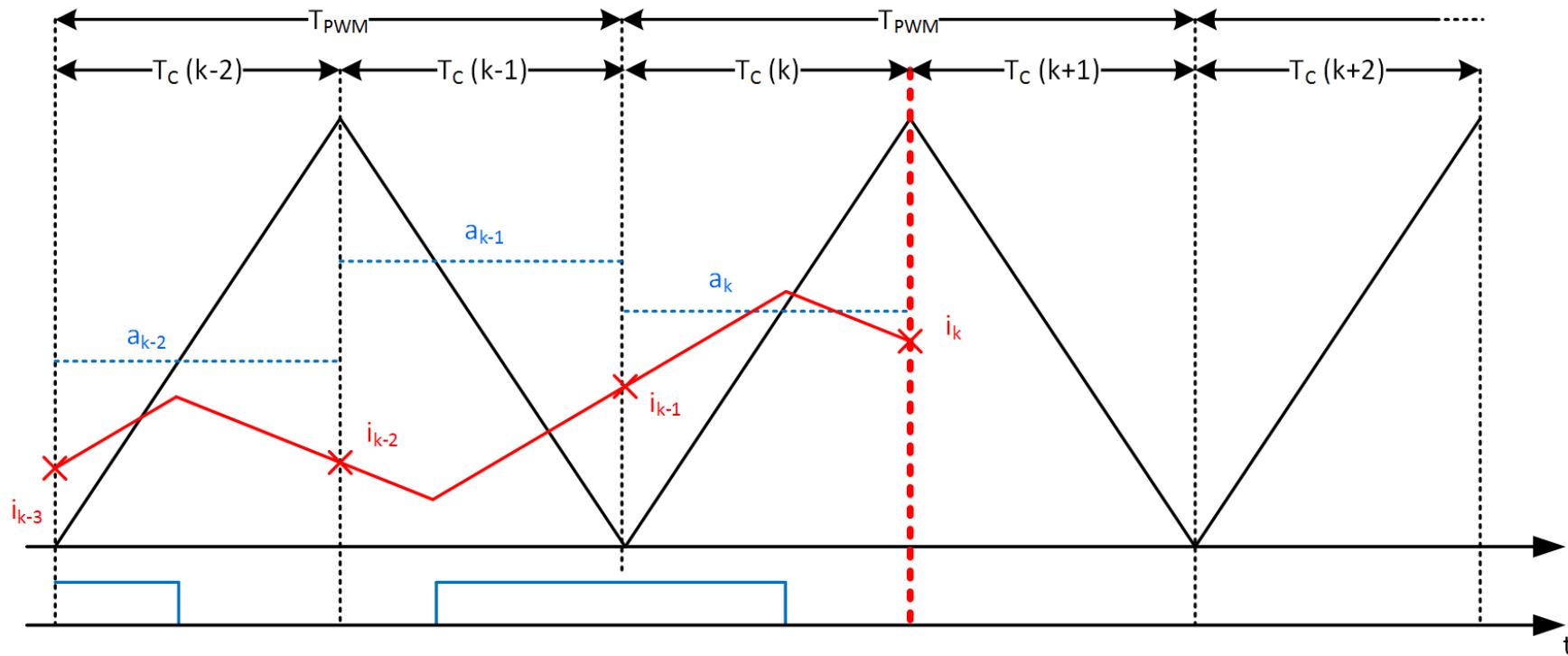
- different current slopes
- ***Different measurement values***

→ **Current slopes can be identified!**



## II. Easy Current Slope Detection (ECSD)

### Derivation of the ECSD-Algorithm



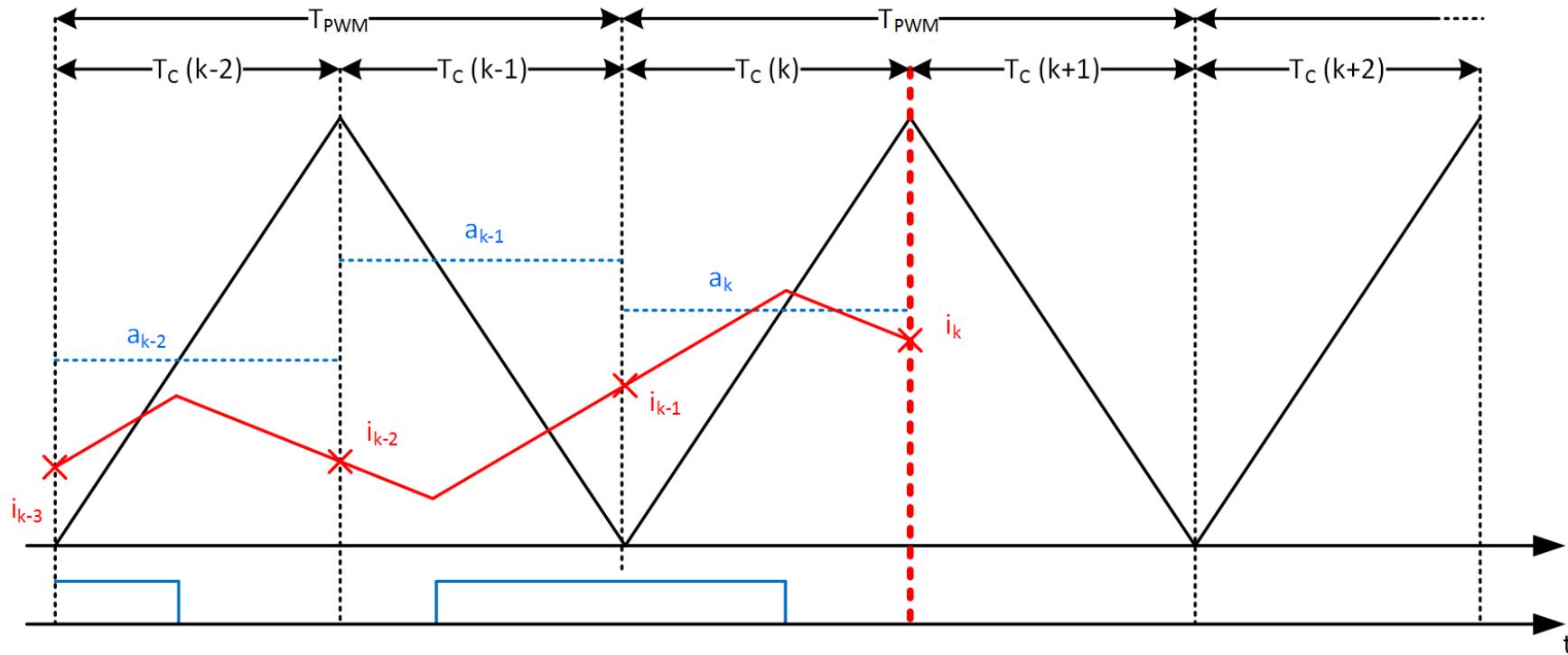
$$i_k = i_{k-1} + a_k \cdot T_C m_a + (1 - a_k) \cdot T_C m_f$$

$$i_{k-1} = i_{k-2} + a_{k-1} \cdot T_C m_a + (1 - a_{k-1}) \cdot T_C m_f$$

2 equations  
2 unknowns

## II. Easy Current Slope Detection (ECSD)

### Derivation of the ECSD-Algorithm

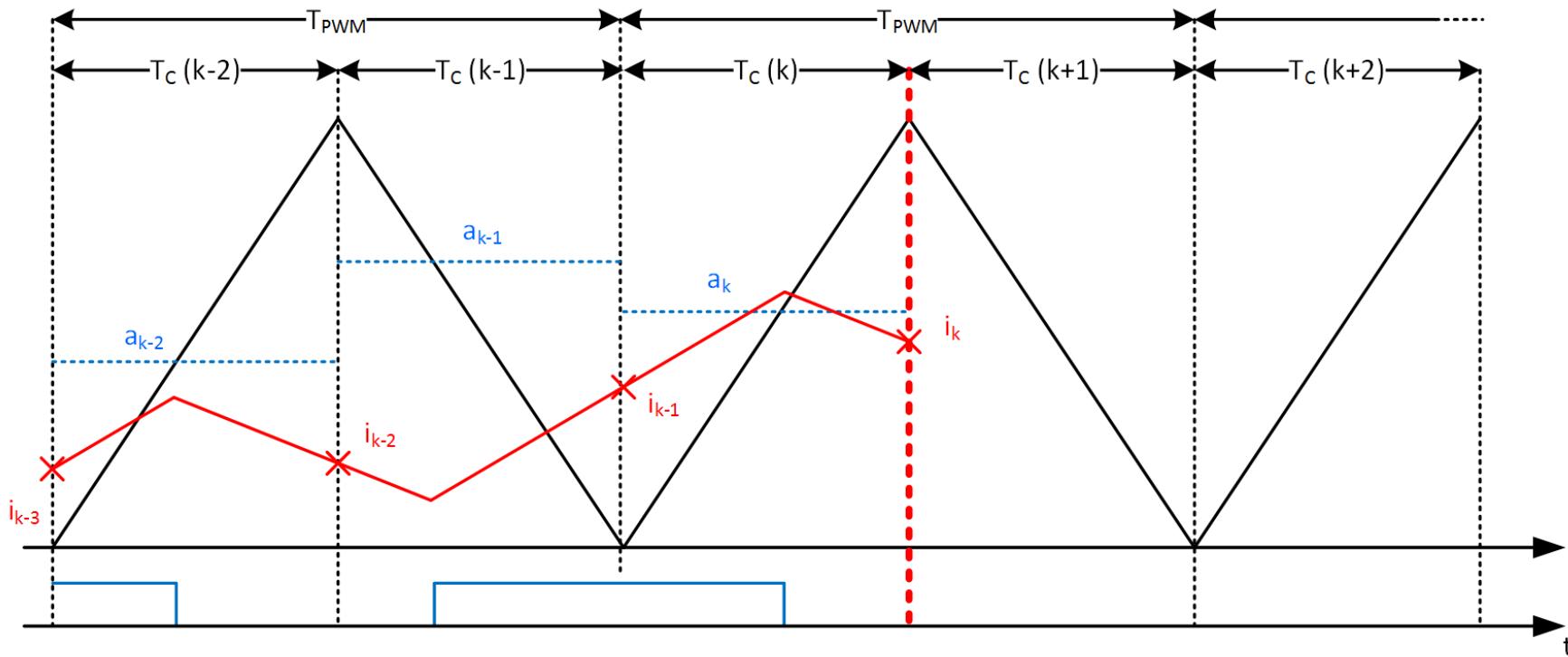


$$m_f = \frac{1}{T_C} \cdot \frac{a_{k-1}(i_{k-1} - i_k) + a_k(i_{k-1} - i_{k-2})}{a_k - a_{k-1}}$$

$$m_a = m_f + \frac{1}{T_C} \cdot \frac{i_{k-2} - 2i_{k-1} + i_k}{a_k - a_{k-1}}$$

## II. Easy Current Slope Detection (ECSD)

### Derivation of the ECSD-Algorithm



$$\Delta i_f = T_C m_f = \frac{a_{k-1}(i_{k-1} - i_k) + a_k(i_{k-1} - i_{k-2})}{a_k - a_{k-1}}$$

$$\Delta i_a = T_C m_a = \Delta i_f + \frac{i_{k-2} - 2i_{k-1} + i_k}{a_k - a_{k-1}}$$

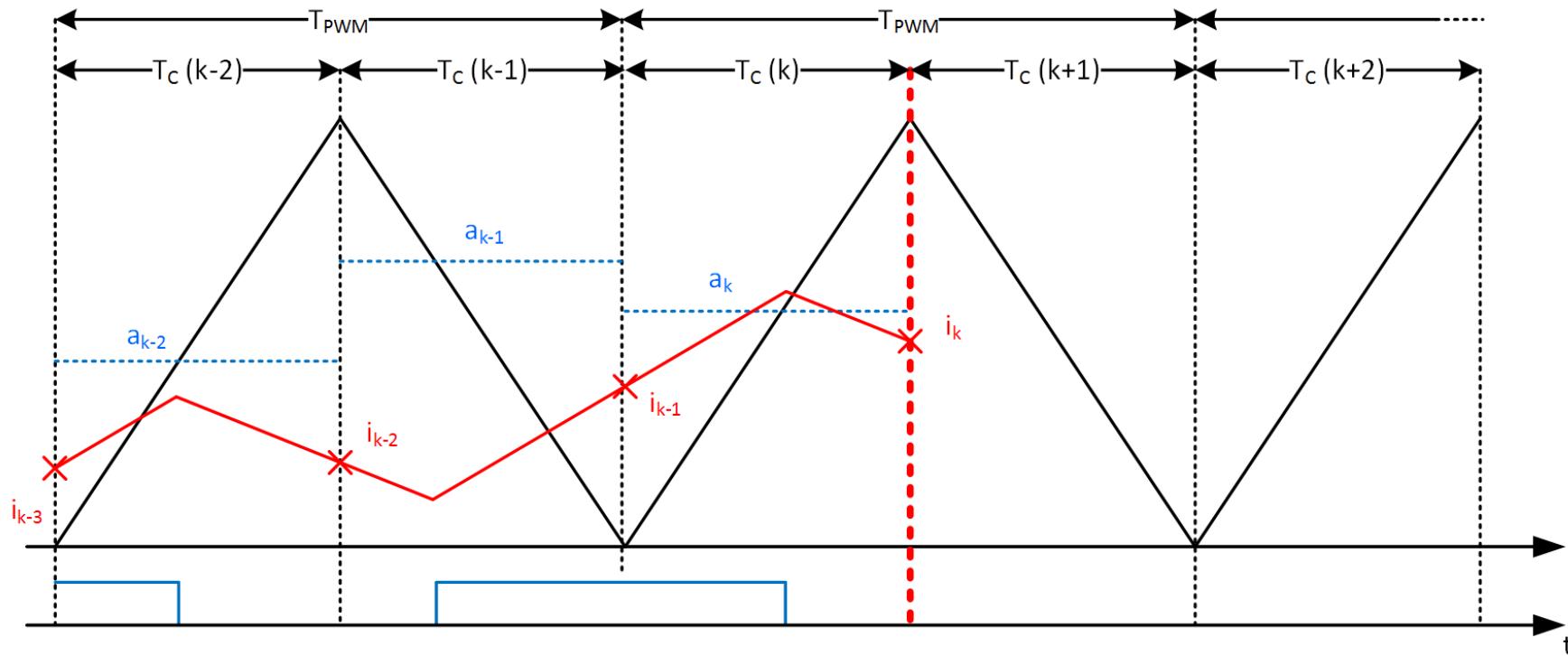
„Current Gradients“:

$$\Delta i_f = T_C \cdot m_f$$

$$\Delta i_a = T_C \cdot m_a$$

## II. Easy Current Slope Detection (ECSD)

### Derivation of the ECSD-Algorithm



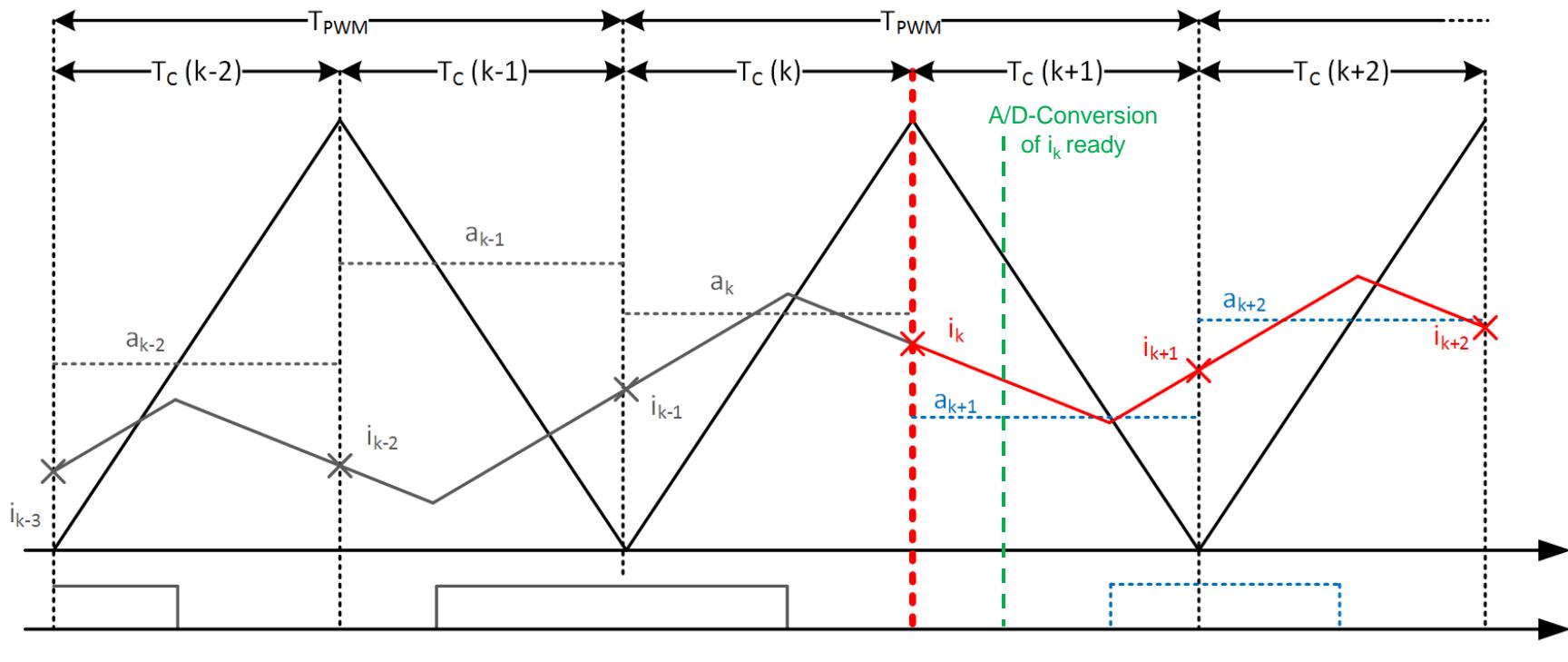
$$\Delta i_f = T_C m_f = \frac{a_{k-1}(i_{k-1} - i_k) + a_k(i_{k-1} - i_{k-2})}{a_k - a_{k-1}}$$

$$\Delta i_a = T_C m_a = \Delta i_f + \frac{i_{k-2} - 2i_{k-1} + i_k}{a_k - a_{k-1}}$$

„Easy Current Slope Detection“

### III. Implementing the ECSD for the DACC

#### Using the Current Gradients for the DACC

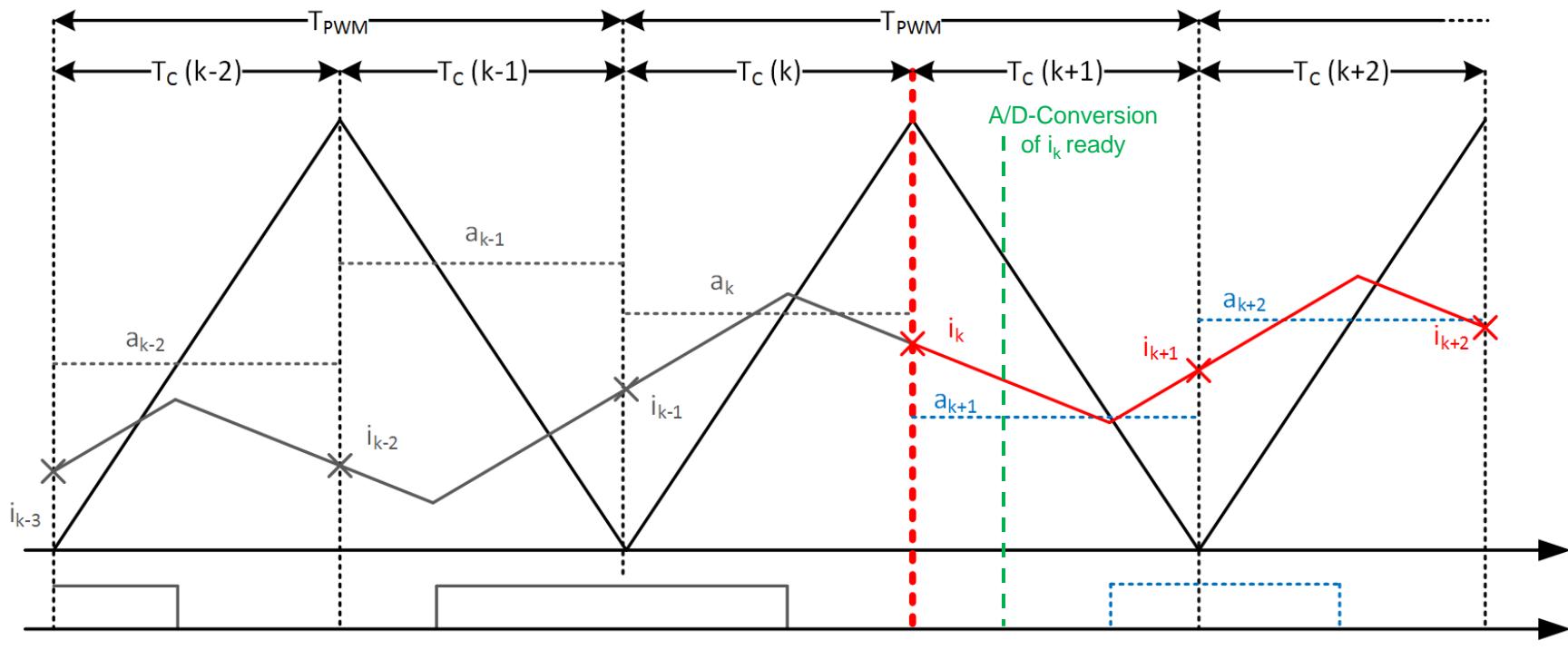


$$i_{k+1} = i_k + a_{k+1} \cdot \Delta i_a + (1 - a_{k+1}) \cdot \Delta i_f$$

$$i_{k+2} = i_{k+1} + a_{k+2} \cdot \Delta i_a + (1 - a_{k+2}) \cdot \Delta i_f$$

### III. Implementing the ECSD for the DACC

#### Using the Current Gradients for the DACC

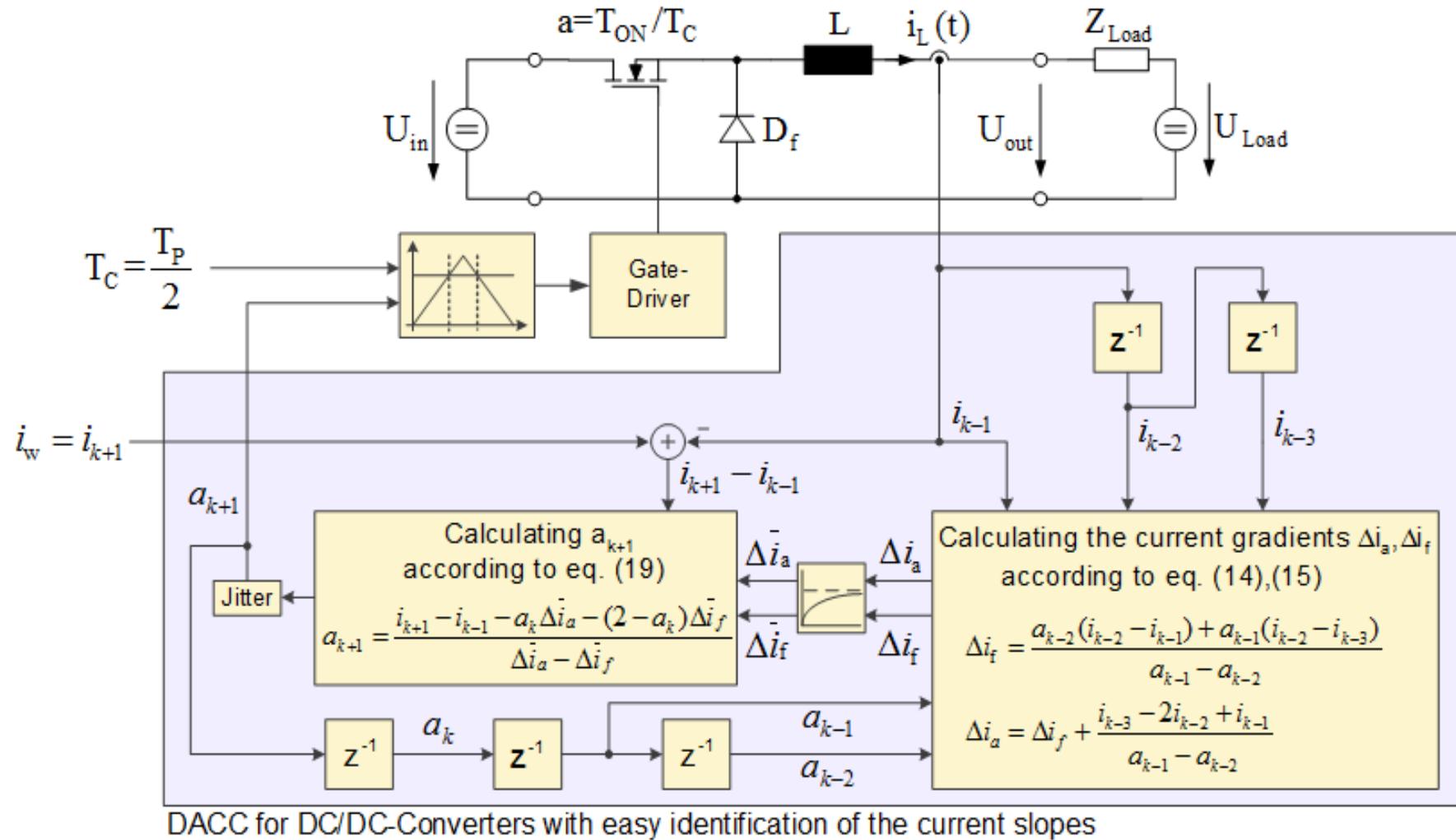


$$a_{k+2} = \frac{i_{k+2} - i_k - a_{k+1} \cdot \Delta i_a - (2 - a_{k+1}) \Delta i_f}{\Delta i_a - \Delta i_f}$$

Direct Adaptive Current Control  
for  
DC/DC-Converters

### III. Implementing the ECSD for the DACC

#### Block Diagram



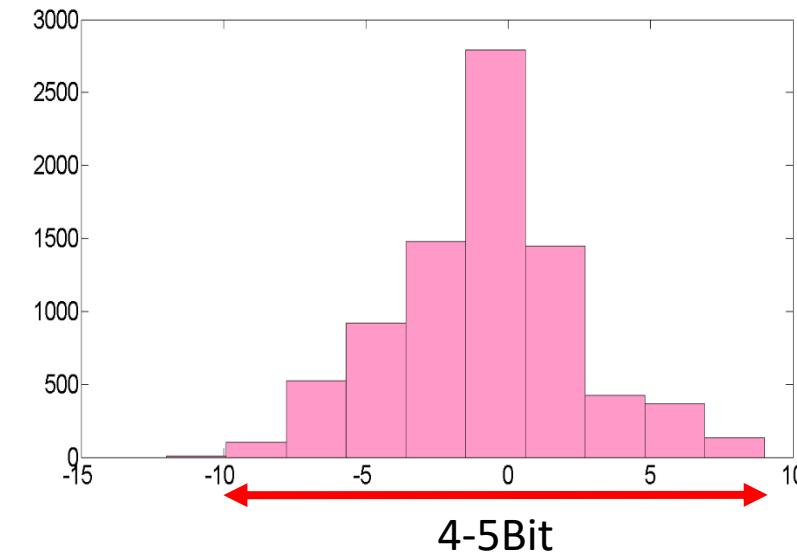
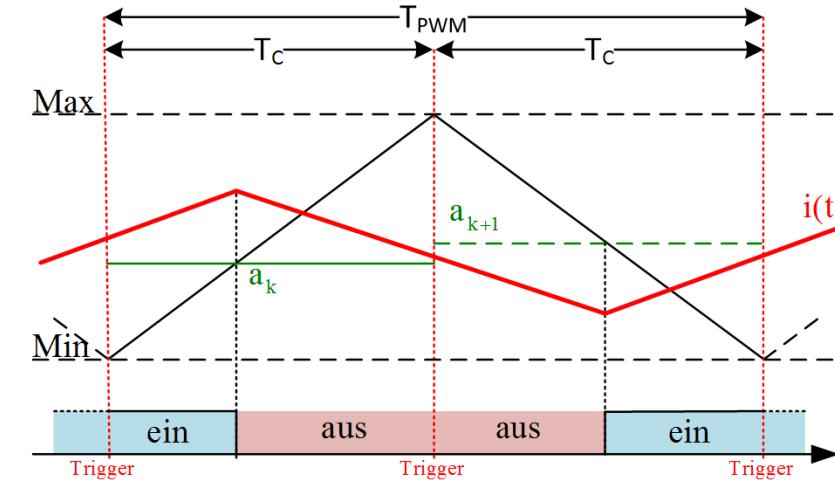
### III. Implementing the ECSD for the DACC

#### Noise & Sensitivity

##### Sensitivity

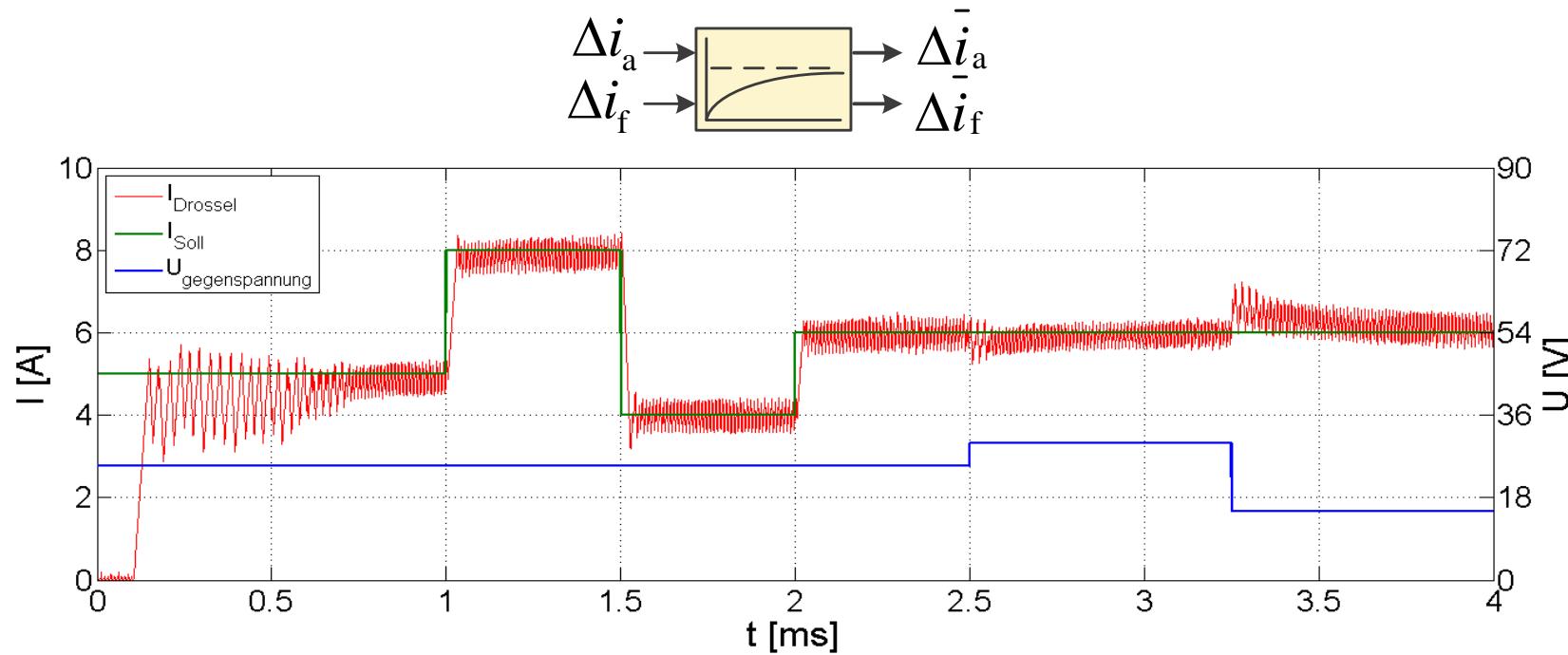
- In steady state operation:  
Almost identical current samples  
→ Difference  $i_k - i_{k-1}$  is very small compared to the measuring range
  
- Critical:
  - Noise of the current signal
  - Noise of the AD-Converter

Original idea of controlling the current with the current slopes, calculated from period to period only did not work due to noise.



### III. Implementing the ECSD for the DACC

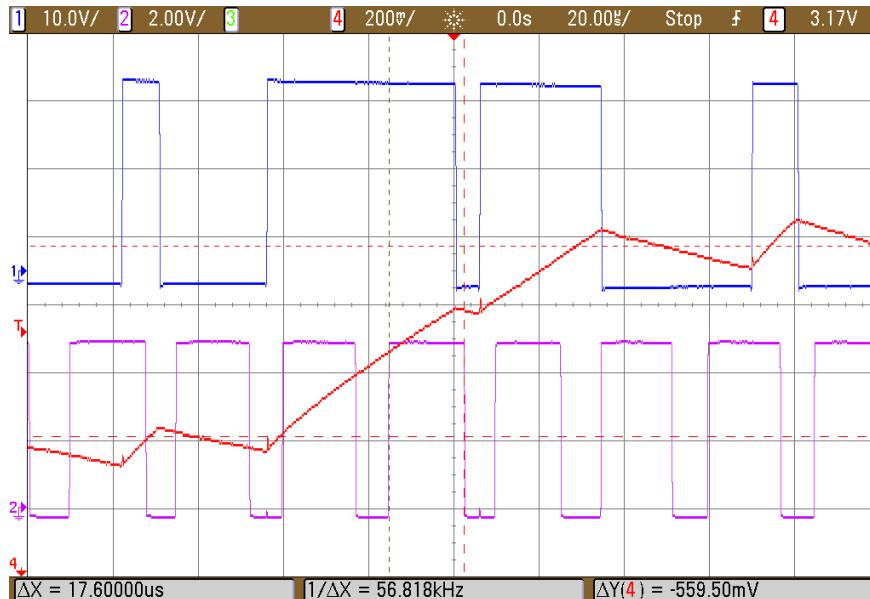
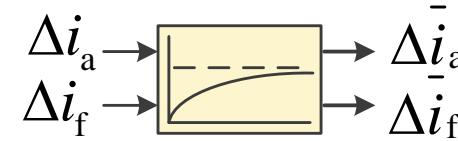
#### DACC with filtered current gradients (DACC-F)



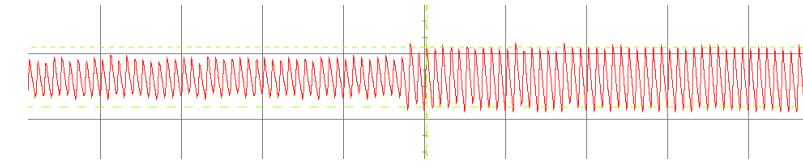
- ✓ only current measurement necessary
- ✓ self-parametrizing
- ✓ dead-beat-characteristics for setpoint steps
- ✗ without voltage feed-forward slow disturbance reaction

### III. Implementing the ECSD for the DACC

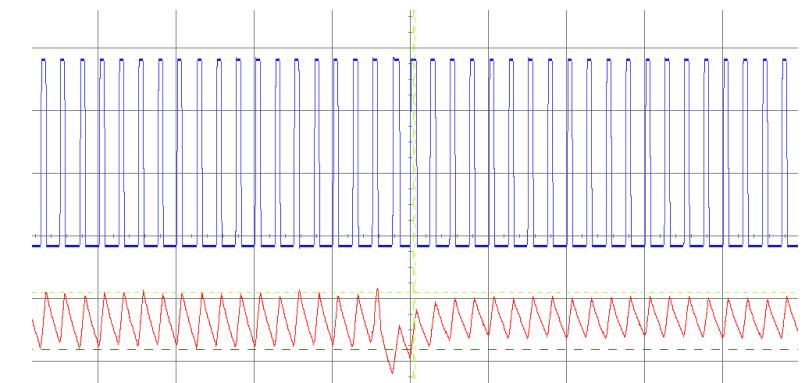
#### DACC with filtered current gradients (DACC-F) – with low-cost microcontroller



Setpoint Step 1A to 3A



Load step: 800nH  $\rightarrow$  500nH



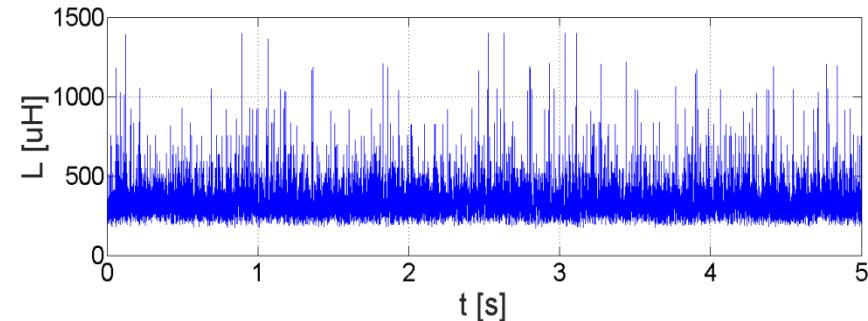
Load step: 600nH  $\rightarrow$  800nH

### III. Implementing the ECSD for the DACC

#### Using the ECSD to calculate L

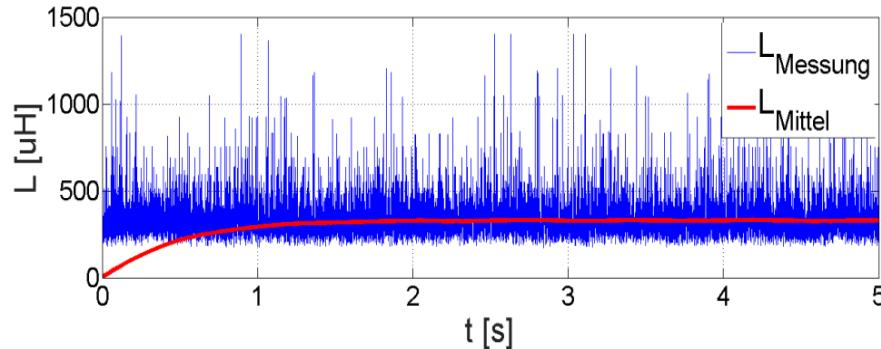
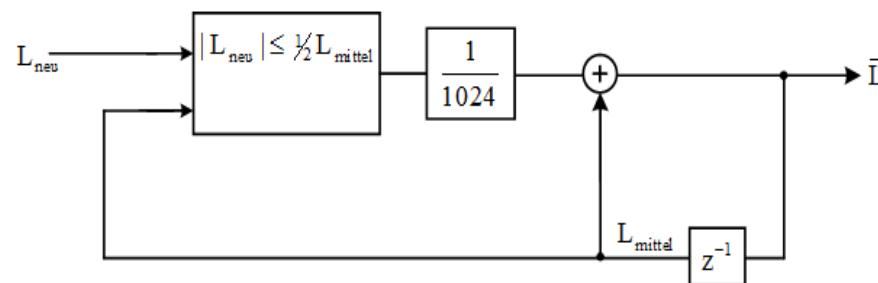
Calculation of L:

$$L = \frac{-U_a \cdot (a_k - a_{k-1})}{a_{k-1} \cdot (i_{k-1} - i_k) + a_k \cdot (i_{k-1} - i_{k-2})}$$



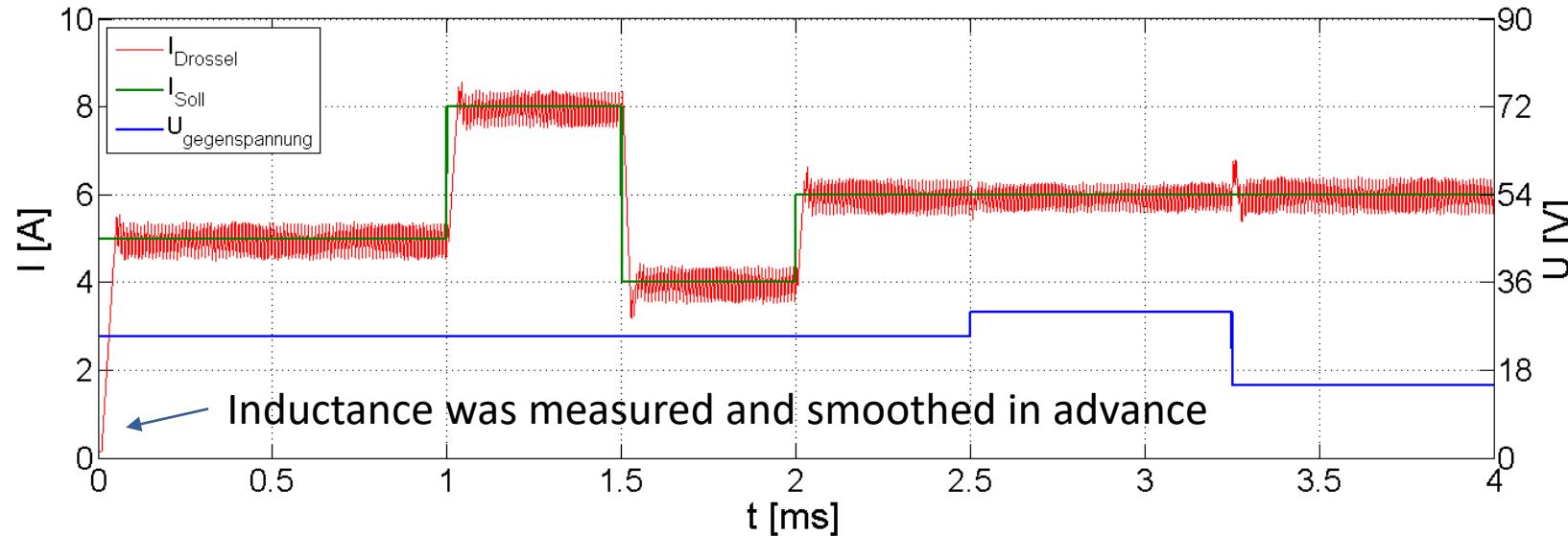
Filtering L:

- Noise is one-sided
- Special Filter



### III. Implementing the ECSD for the DACC

#### MPC with filtered L (DACC-MPC)



- ✓ Excellent control quality and good disturbance reaction
- ✗ Additional voltage-measurement necessary
- ✗ Smoothing out L takes relatively long ( $\sim 1s$ )

## IV. Conclusion

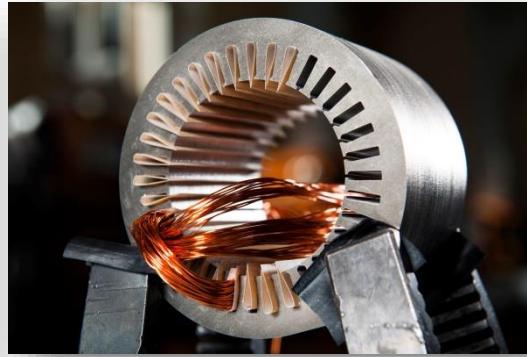
### 1. Easy & efficient algorithm to identify the current slopes (ECSD)

- ✓ simple & low cost
- ✓ easy to implement
- ✓ no external hardware needed
- ✓ just two samples per period at times where switching noise is minimal
- ✓ perfect for high switching frequency applications

### 2. Combination of the ECSD with the Direct Adaptive Current Control (DACC)

- ✓ Enables the DACC in high switching frequency SMPS
- ✓ Realized as DACC-F and MPC
- ✓ Deadbeat characteristics in setpoint steps
- ✓ despite jitter good steady state accuracy
- ✓ self-initializing & low cost
- ✓ inherent stable

# Thank you for your attention!



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